

1.1 MHz, 3A Synchronous Step-Down Converter

FEATURES

- High Efficiency: Up to 96%
- 1.1MHz Constant Frequency Operation
- 3A Output Current
- No Schottky Diode Required
- 2.5V to 6.0V Input Voltage Range
- Output Voltage as Low as 0.6V
- PFM Mode for High Efficiency in Light Load
- 100% Duty Cycle in Dropout Operation
- Low Quiescent Current: 40µA
- Short Circuit Protection
- Thermal Fault Protection
- Inrush Current Limit and Soft Start
- QFN2X2-6 package

APPLICATIONS

- Cellular and Smart Phones
- Microprocessors and DSP Core Supplies
- Wireless and DSL Modems
- PDA
- MP3 Player
- Digital Still and Video Cameras
- Portable Instruments

GENERAL DESCRIPTION

The GX3430 are high-efficiency, high frequency synchronous step-down DC-DC regulator ICs capable of delivering up to 3A output currents. The GX3430 can operateover a wide input voltage range from 2.5V to6.0V and integrate main switch and synchronous switch with very low RDS(ON)to minimize the conduction loss.It is idealfor powering portable equipmentthat runsfrom a single cell Lithium-Ion (Li+)battery. The output voltage can be regulated s lowas 0.6V. The GX3430 can also runat100% duty cycle for low dropout operation, extending battery life in portable system. This device offers two operation modes, PWM control and PFM Mode switchingcontrol, which allows a high efficiency overthe wider range of the load. The GX3430 isoffered in a low profile 6-pin, thin QFN2X2package, and is availablein an adjustableversion.

TYPICAL APPLICATION

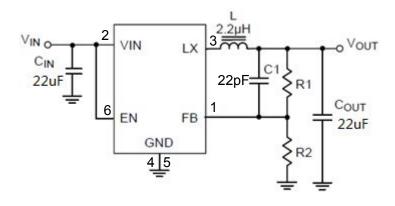
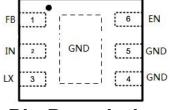


Figure 1: High Efficiency Step-down Converter

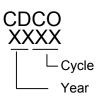


Package/Order Information



Pin Description

Part Number	Top Mark	Temp Range
GX3430	CDCO XXXX	-40°C to +85°C



PIN	NAME	FUNCTION(QFN)
1	FB	Feedback Input Pin. Connect FB to the center point of the external resistor divider. The feedback threshold voltage is 0.6V.
2	IN	Supply Input Pin. Must be closely decoupled to GND, Pin 4, with a 22 μ F or greater ceramic capacitor.
3	LX	Power Switch Output. It is the Switch note connection to Inductor. This pin connects to the drains of the internal P-CH and N-CH MOSFET switches.
4	GND	Ground
5	GND	Ground
6	EN	Regulator Enable control input. Drive EN above 1.5V to turn on the part. Drive EN below 0.3V to turn it off. In shutdown, all functions are disabled drawing <1µA supply current. Do not leave EN floating.

Functional Block Diagram

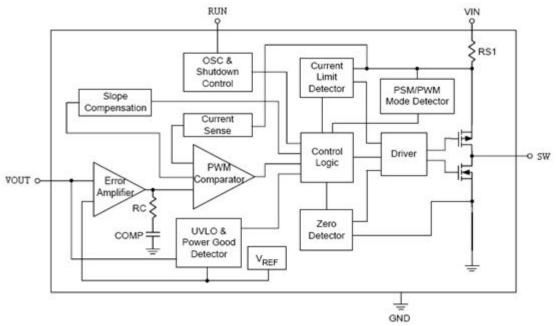


Figure 2. GX3430 Block Diagram



Absolute Maximum Rating (Note 1)

Input Supply Voltage	-0.3V to +7.5V
EN, V _{FB} Voltages0.3V	to V _{IN} +0.3V
SW, Vout Voltages0.3V	to V _{IN} +0.3V
Peak SW Sink and Source Current	t3.5A

Operating Temperature Range-40°C to +85°C Junction Temperature ^(Note 2).....+125°C Storage Temperature Range. -65°C to +150°C Lead Temperature (Soldering, 10S).....+300°C

Electrical Characteristics (Note 3)

(V_{IN}=V_{RUN}=3.6V, V_{OUT}=1.8V, T_A = 25°C, unless otherwise noted.)

Parameter	Conditions	MIN	TYP	MAX	unit
Input Voltage Range		2.5		6.0	V
UVLO Threshold			2.4		V
OVP Threshold			6.0		V
Input DC Supply Current PWM Mode PFM Mode Shutdown Mode	Vout = 90%, Iload=0mA Vout = 105%, Iload=0mA V _{EN} = 0V, V _{IN} =4.2V		150 40 0.1	300 70 1.0	μΑ μΑ μΑ μΑ
Regulated Feedback Voltage	T _A = 25°C	0.588	0.600	0.612	V
	$T_A = 0^{\circ}C \le T_A \le 85^{\circ}C$	0.586	0.600	0.613	V
	$T_A = -40^{\circ}C \le T_A \le 85^{\circ}C$	0.585	0.600	0.615	V
Reference Voltage Line Regulation	Vin=2.5V to 5.5V		0.04	0.40	%/V
Output Voltage Line Regulation	V _{IN} = 2.5V to 5.5V		0.04	0.4	%
Output Voltage Load Regulation			0.5		%
Oscillation Frequency			1.1		MHz
On Resistance of PMOS	I _{sw} =100mA		0.08		Ω
ON Resistance of NMOS	I _{sw} =-100mA		0.06		Ω
Peak Current Limit	V _{IN} = 3V, Vout=90%	3			А
EN Threshold		1.5			V
EN Leakage Current			±0.01	±1.0	μA
SW Leakage Current	V _{EN} =0V,V _{IN} =Vsw=5V		±0.01	±1.0	μA
Thermal Shutdown			160		°C

Note 1: Absolute Maximum Ratings are those values beyond which the life of a device may be impaired.

Note 2: T_J is calculated from the ambient temperature T_A and power dissipation P_D according to the following formula: $T_J = T_A + (P_D) \times (250^{\circ}C/W).$

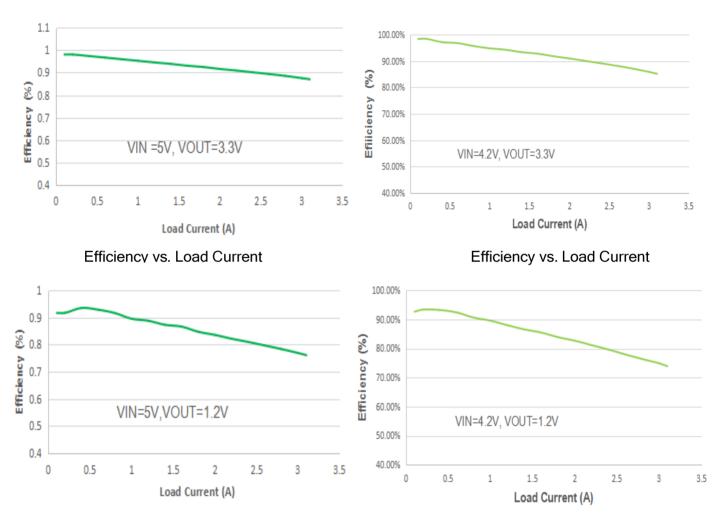
Note3: 100% production test at +25°C. Specifications over the temperature range are guaranteed by design and characterization.

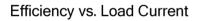


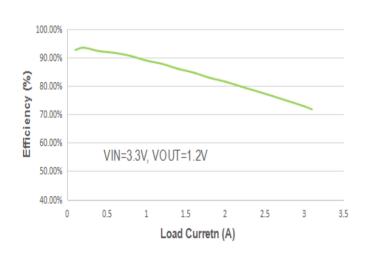
Typical Performance Characteristics

Efficiency vs. Load Current











FUNCTIONAL DESCRIPTION

GX3430 is a synchronous buck regulator IC that integrates the PWM/PFM control, top and bottom switches on the same die to minimize the switching transition loss and conduction loss. With ultra-low RDS(ON) power switches and proprietary PWM control, this regulator IC can achieve the highest efficiency and the highest switch frequency simultaneously to minimize the external inductor and capacitor size, and thus achieving the minimum solution footpr

TheGX3430 requires only three external power components (Cin, Cout and L). The adjustable version can be programmed with external feedback to any voltage, ranging from 0.6V to the input voltage. At dropout, the converter duty cycle increases to 100% and the output voltage tracks the input voltage minus the Rdson drop of the high-side MOSFET.

The internal error amplifier and compensation provides excellent transient response, load, and line regulation. Soft start function prevents input inrush current and output overshoot during start up.

Setting the Output Voltage

Figure 1 above shows the basic application circuit with GX3430adjustable output version. The external resistor sets the output voltage according to the following equation:

$$V_{OUT} = 0.6V \times \left(1 + \frac{R1}{R2}\right)$$

Table 1 Resistor select for output voltage setting

V _{OUT}	R2	R1
1.2V	100k	100k
1.5V	100k	150k
1.8V	100k	200k
2.5V	100k	316k
3.3V	100k	453k

Note: R2 Resistor must be lower than 100K.

APPLICATIONS INFORMATION

Setting the Output Voltage

The external resistor divider is used to set the output voltage (see Typical Application on page 1). The feedback resistor R1 also sets the feedback loop bandwidth with the internal compensation capacitor. Choose R1 to be around $100k\Omega$ for optimal transient response. R2 is then given by:

$$R_2 = \frac{R_1}{V_{out} / V_{FB} - 1}$$

Inductor Selection

For most designs, the GX3430 operates with inductors of 1μ H to 4.7μ H. Low inductance values are physically smaller but require faster switching, which results in some efficiency loss. The inductor value can be derived from the following equation:



$$L = \frac{V_{OUT} \times (V_{IN} - V_{OUT})}{V_{IN} \times \Delta I_L \times f_{OSC}}$$

Where ΔI_L is inductor Ripple Current. Large value inductors result in lower ripple current and small value inductors result in high ripple current. For optimum voltage-positioning load transients, choose an inductor with DC series resistance in the 50m Ω to 150m Ω range.

Input Capacitor Selection

With the maximum load current at 3A, the maximum ripple current through input capacitor is about 0.6Arms. A typical X7R or better grade ceramic capacitor with 6V rating and greater than 10uF capacitance can handle this ripple current well. To minimize the potential noise problem, place this ceramic capacitor really close to the IN and GND pins. Care should be taken to minimize the loop area formed by CIN, and IN/GND pins.

Output Capacitor Selection

The output capacitor is required to keep the output voltage ripple small and to ensure regulation loop stability. The output capacitor must have low impedance at the switching frequency. Ceramic capacitors with X5R or X7R dielectrics are recommended due to their low ESR and high ripple current ratings. The output ripple V_{OUT} is determined by:

$$\Delta V_{OUT} \leq \frac{V_{OUT} \times (V_{IN} - V_{OUT})}{V_{IN} \times f_{OSC} \times L} \times \left(ESR + \frac{1}{8 \times f_{osc} \times C3} \right)$$

A 10µF ceramic can satisfy most applications.

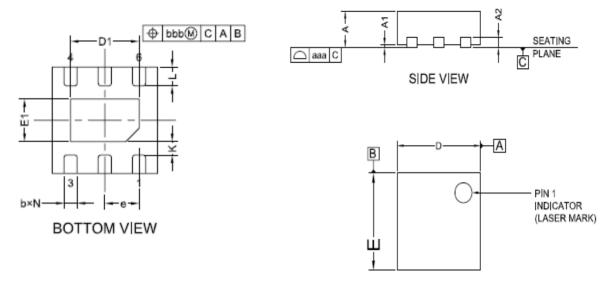
PC Board Layout Checklist

When laying out the printed circuit board, the following checking should be used to ensure proper operation of theGX3430. Check the following in your layout:

- 1. The power traces, consisting of the GND trace, the SW trace and the VIN trace should be kept short, direct and wide.
- Does the (+) plates of Cin connect to VIN as closely as possible? This capacitor provides the AC current to the internal power MOSFETs.
- 3. Keep the switching node, SW, away from the sensitive VOUT node.
- 4. Keep the (-) plates of Cin and Cout as close as possible.



QFN2x2-6 Package Outline Drawing



TOP VIEW

COMMON DIMENSIONS (UNITS OF MEASURE=MILLIMETER)

SYMBOL	MIN	TYP	MAX	
А	0.70	0,75	0.80	
A1	0.00	0.02	0.05	
A2	0.203			
b	0.20	0.25	0.30	
D	1,95	2.00	2.05	
D1	1.20	1.30	1.40	
E	1.95	2.00	2.05	
E1	0.70	0.80	0.90	
е	0.65BSC			
L	0.30	0.35	0,40	
к	0.20MIN			
N	6			
aaa	0.08			
bbb	0.10			